

# (RF) Media-based Modulation

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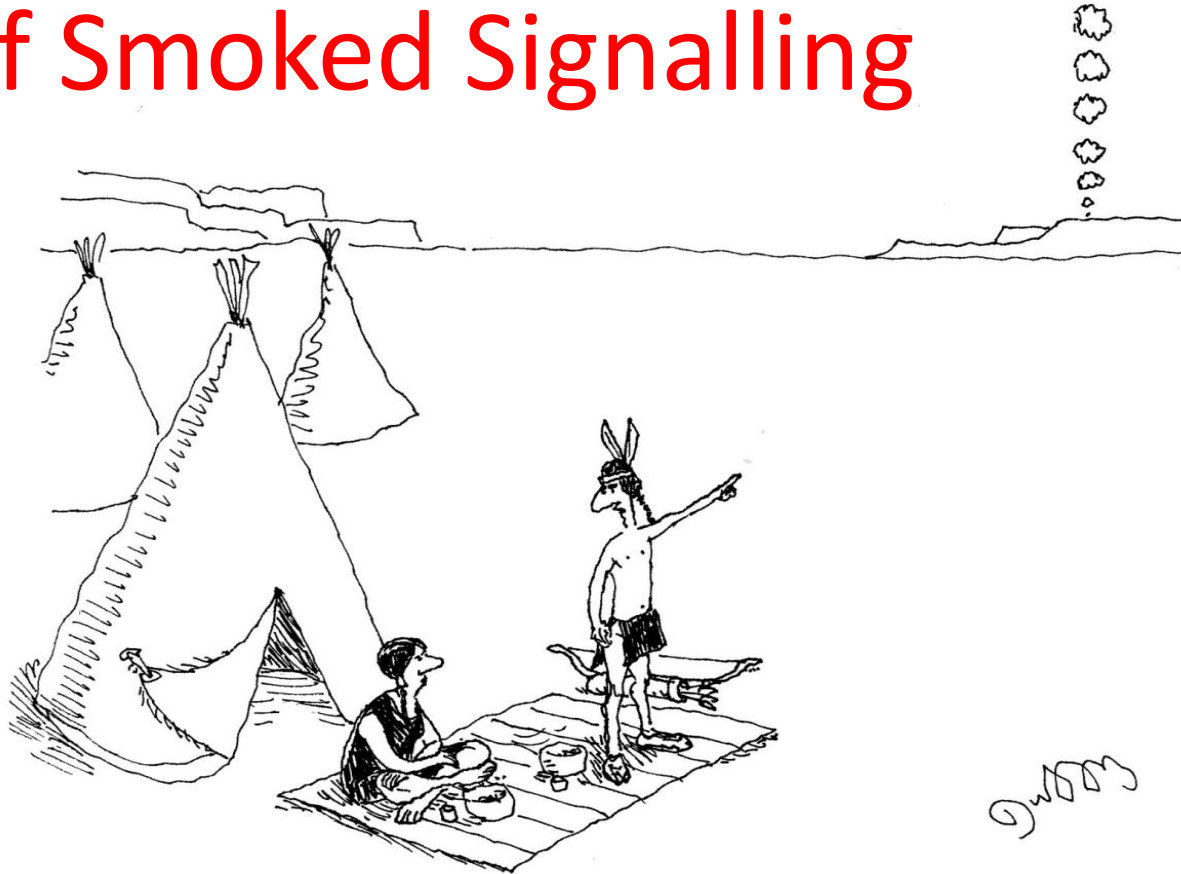
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# Basic Idea:

## Think of Smoked Signalling



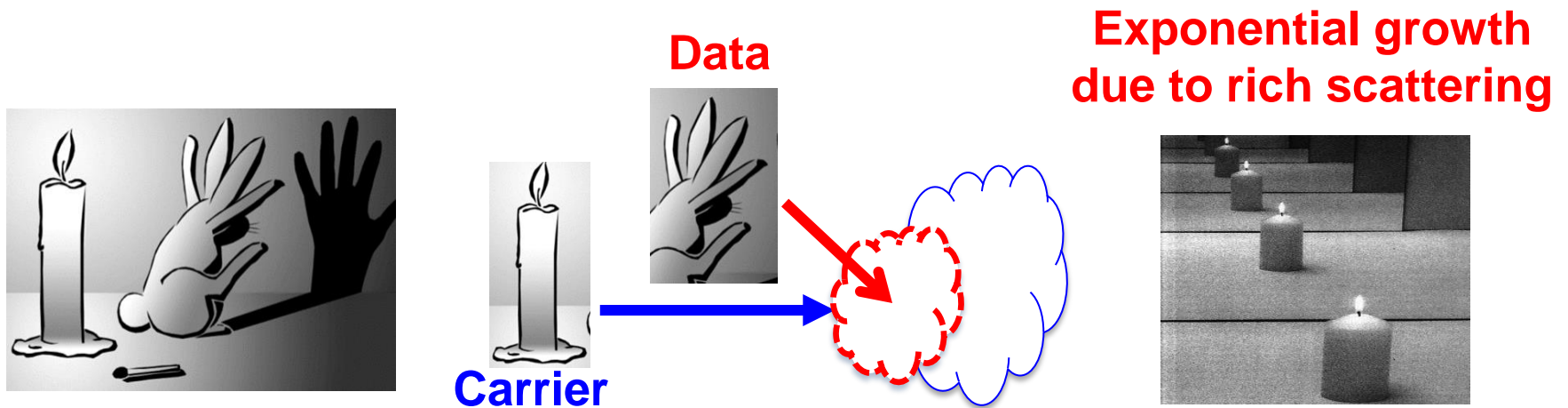
"SORRY, BUT I REALLY NEED TO TAKE THIS."

--- Message is formed outside TX antenna.

--- Message is encoded into pseudorandom channel states.



# Media-based Wireless



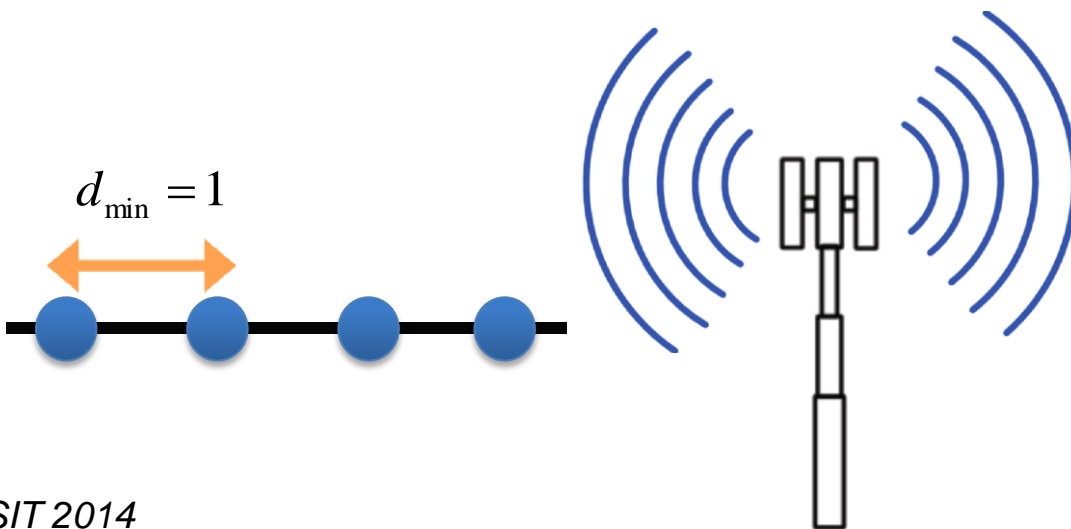
- Keep the source shining and change the transmission medium to embed data.

# Example

- Have two beams of gain **0.5** and **1.5** and have to use only one of them  
-Do not know which beam has higher gain.
- Want to form a 4 points 1-D constellation

**Option 1:** *Select one beam at random and use a 4 PAM constellation*

50% chance of constellation of having a constellation with  $d_{\min} = 1/\sqrt{2}$



**Energy spent: 5/4**

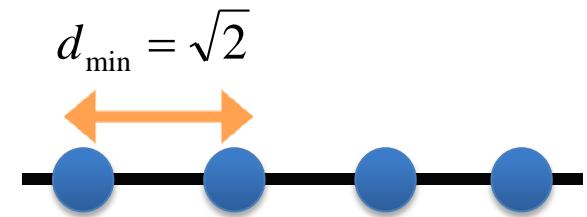
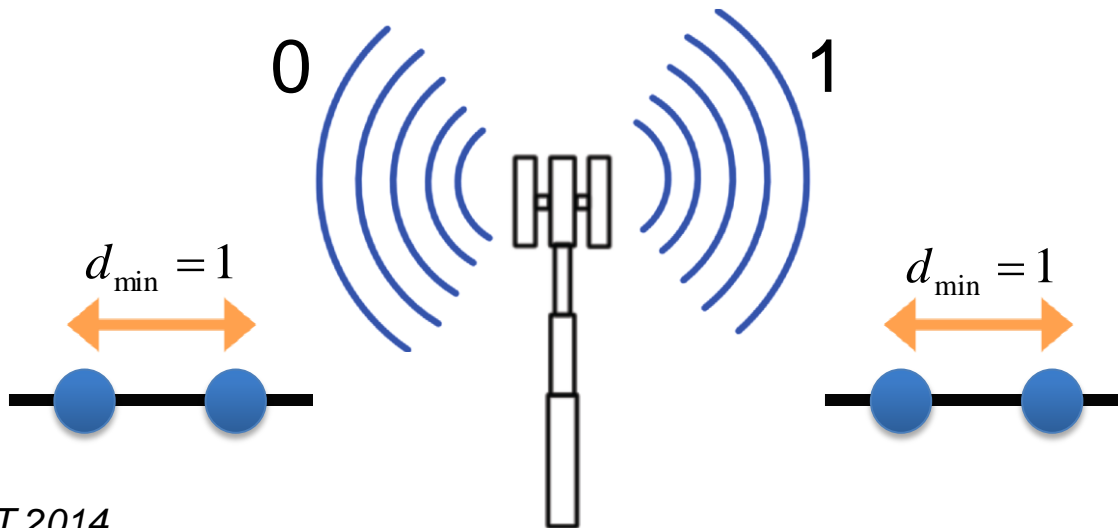
$$d_{\min} = 1/\sqrt{2}$$



# Example

- Have two beams of gain **0.5** and **1.5** and have to use only one of them  
-Do not know which beam has higher gain.
- Want to form a 4 points 1-D constellation

**Option 2:** Select one bits of information to select one of the beams, and modulate a BPSK



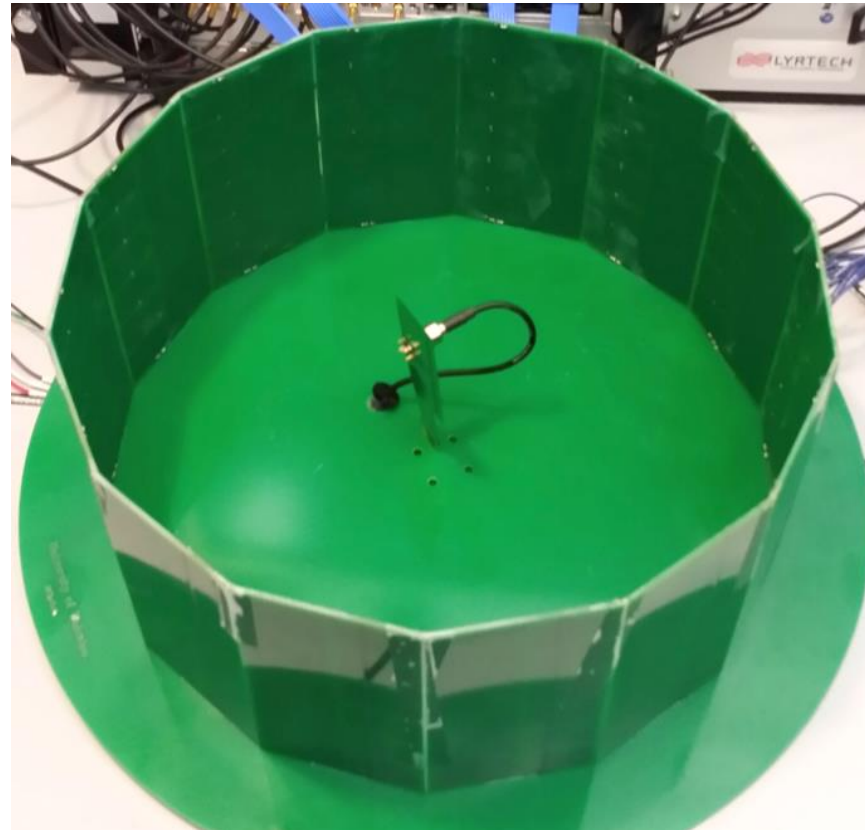
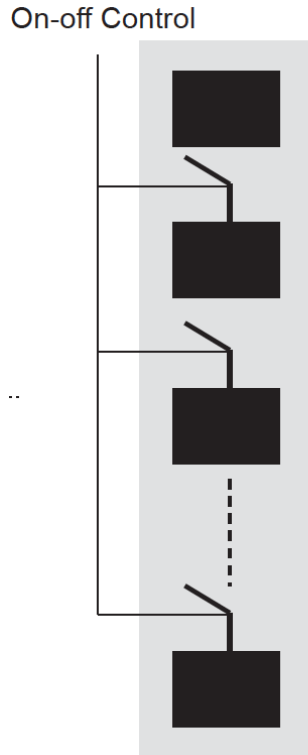
**Energy spent: 1/4**

$$d_{\min} = \sqrt{2}$$

**10 Times  
Energy Saving**



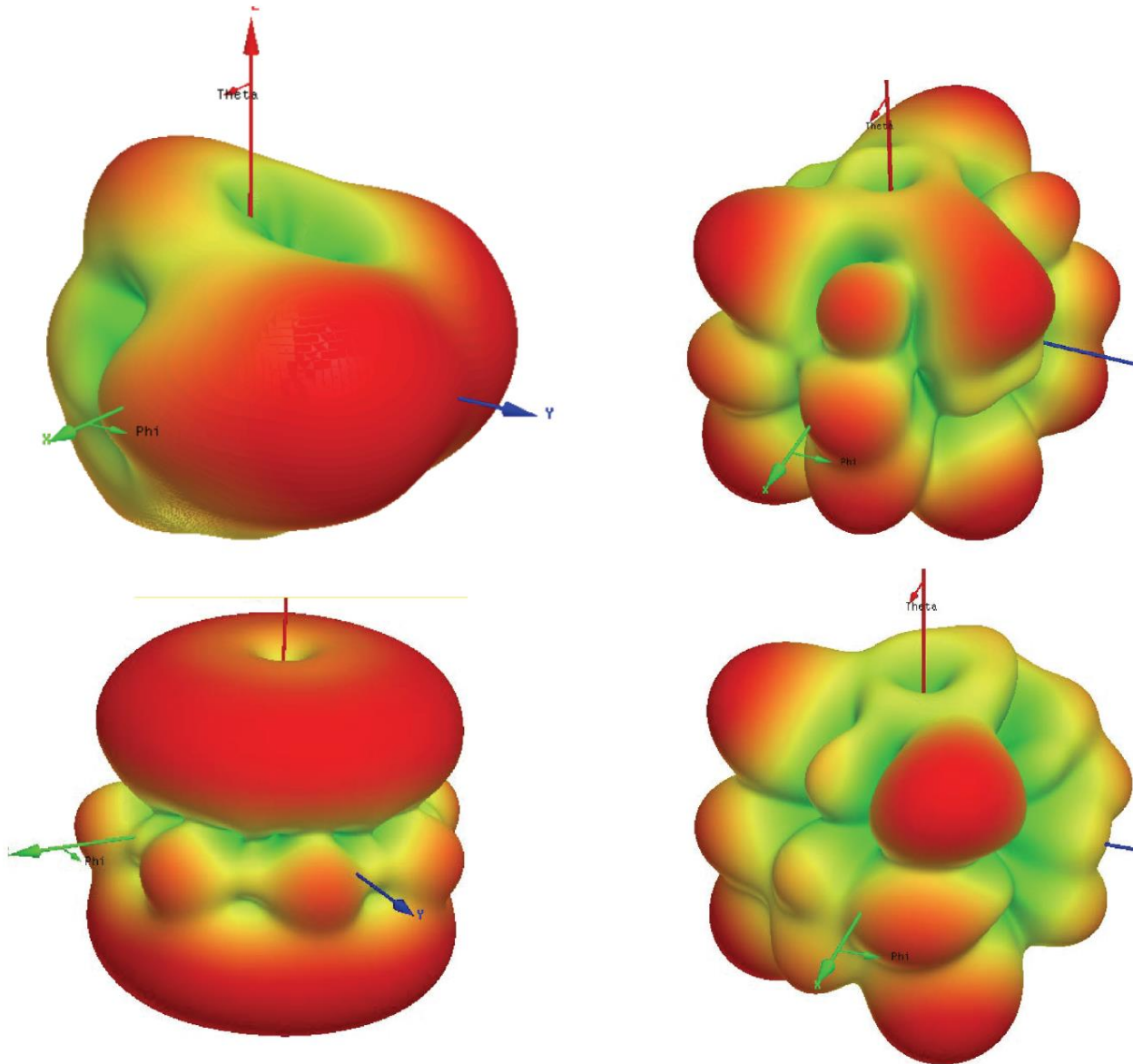
# How to Change the Channel State?



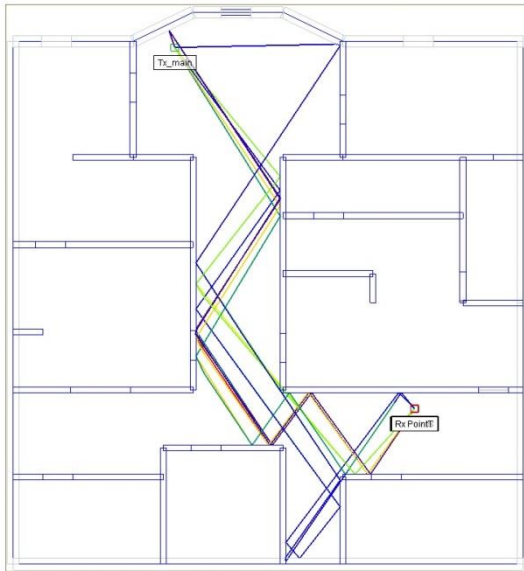
14 RF Mirrors  $\rightarrow$   $2^{14}$  channel states  $\rightarrow$  Modulate 14 bits



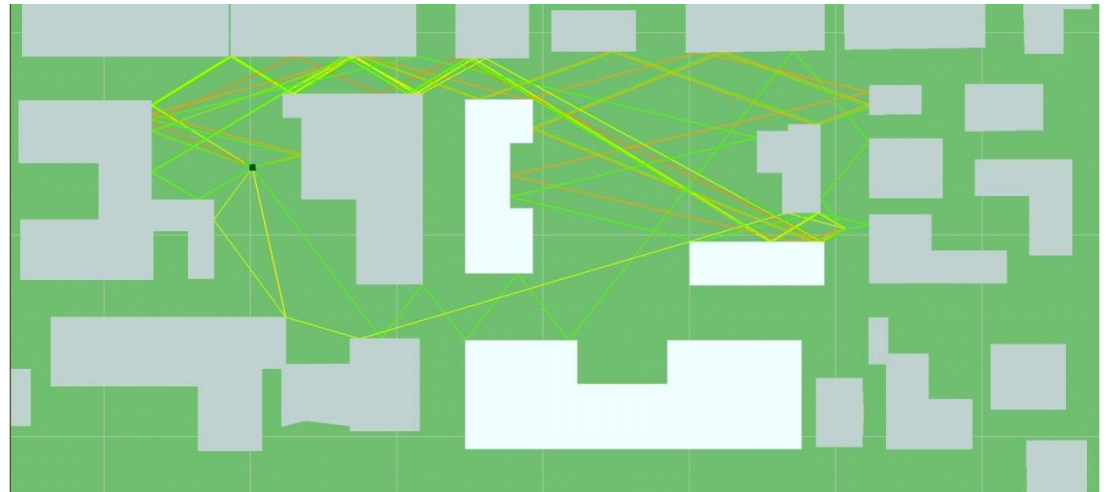
# Examples of Antenna Patterns



# Propagation Environments



Indoor Model  
(residential with dry-walls)

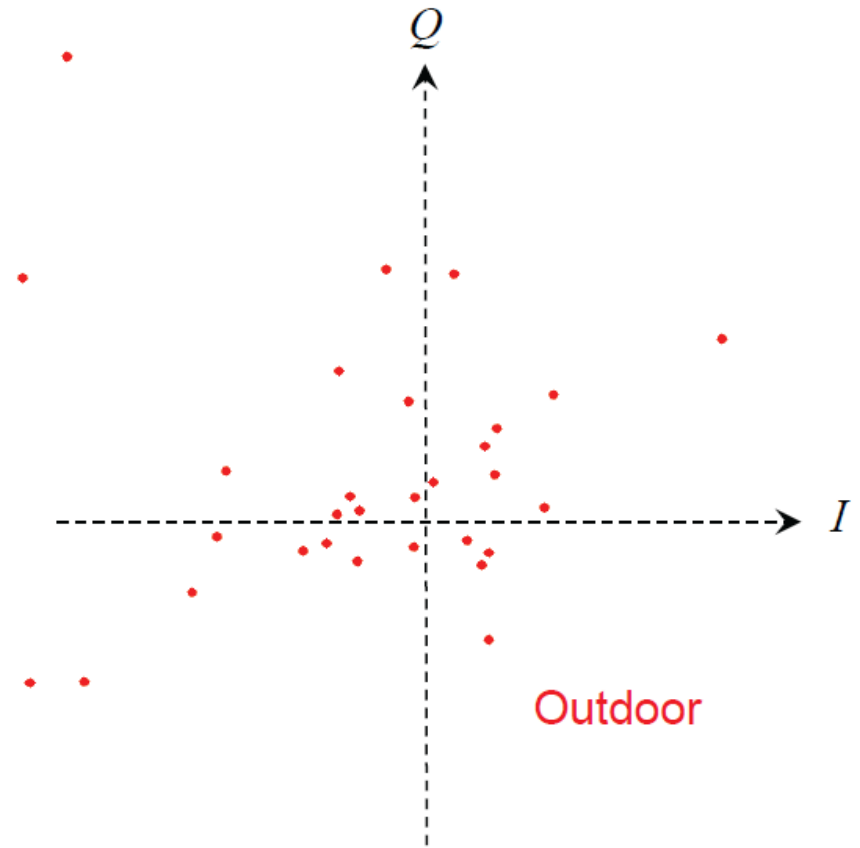
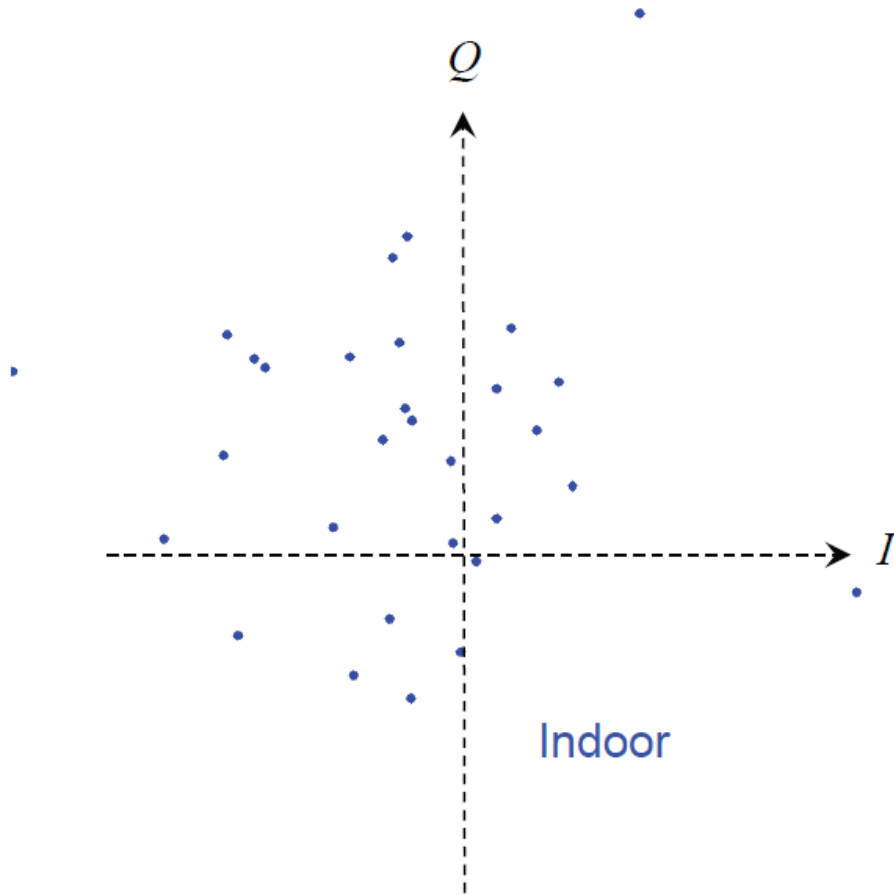


Outdoor Model  
(down-town Ottawa)

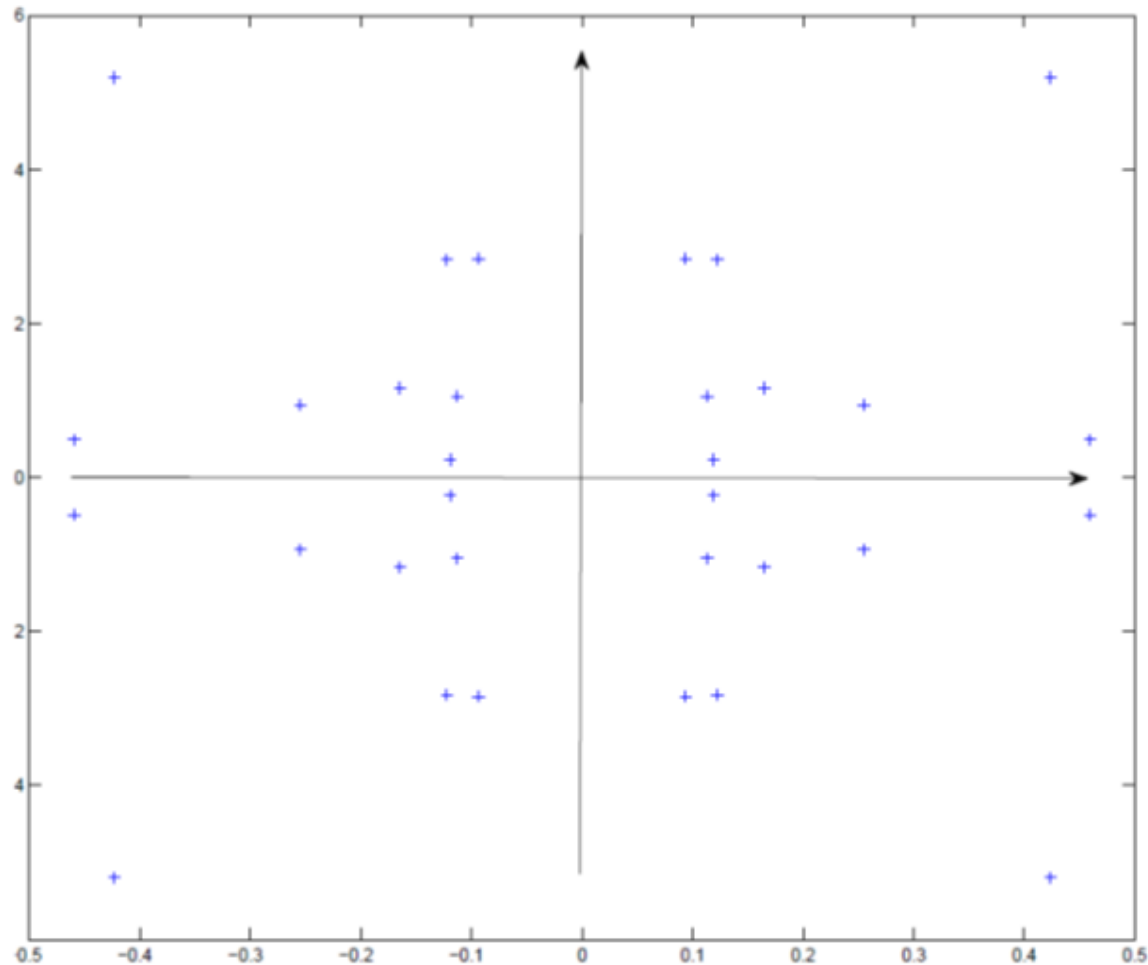




# Examples of Resulting Constellations (without symmetrization)



# Examples of Resulting Constellations (with symmetrization)

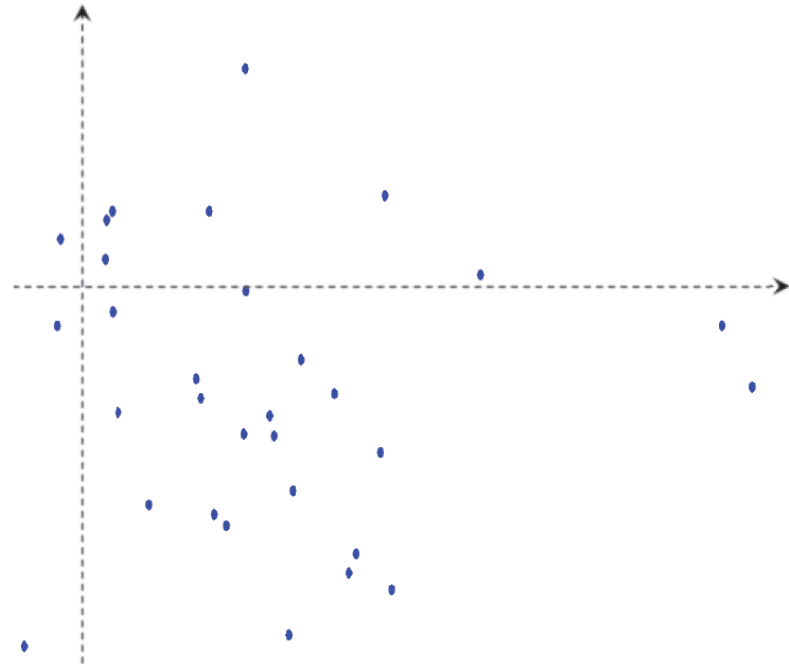
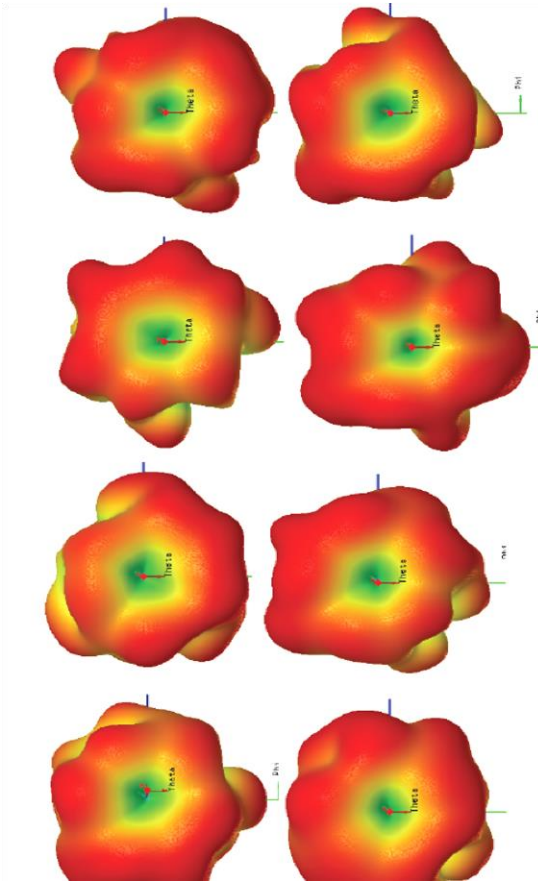


# Some Remarks

- Rich scattering environment:
  - Slightest perturbation in the environment causes independent outcomes.
- **Should not be confused with RF beam-forming using parasitic elements.**
  - RF beam-forming aims at focusing energy.
  - Media-based relies on additive information over receive antennas to increase rate, and on randomness of constellation to combat slow fading.



# What about line of sight

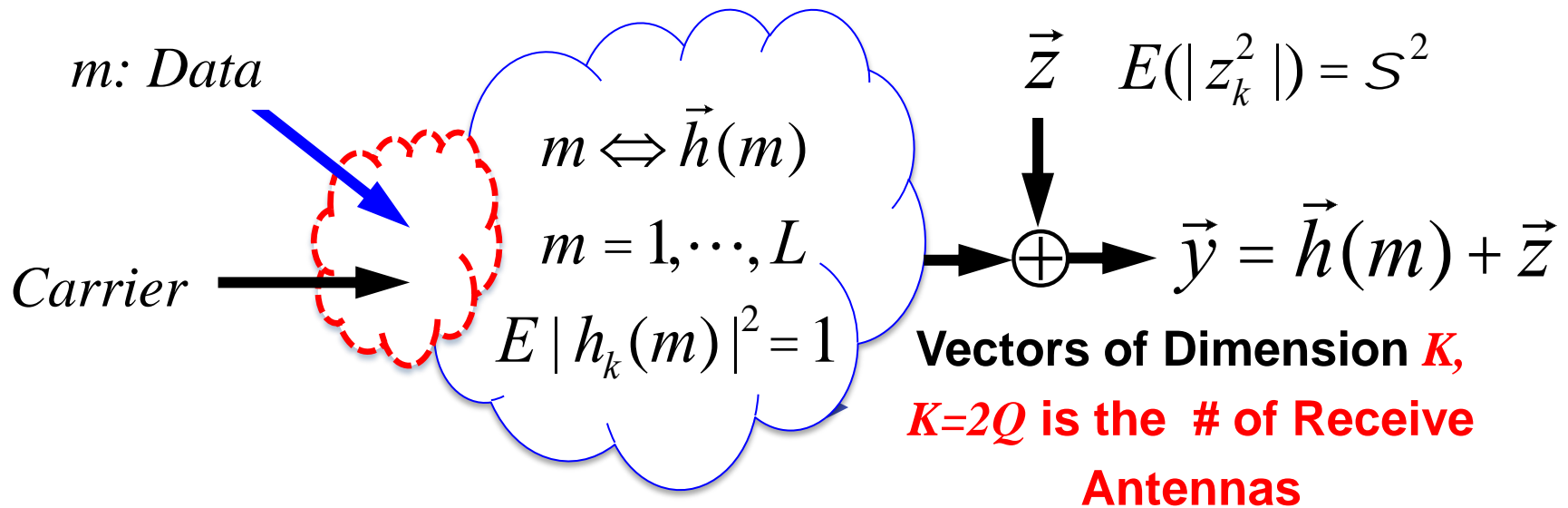


# Media-based vs. (legacy) Source-based

- Main idea:
  - Embed the information in the variation of the RF channel external to the antenna.
- Benefits vs. (legacy) source-based wireless:
  - Additive information over multiple receive antennas (similar to MIMO) with the advantages of:
    - Using a single transmit antenna
    - Independence of noise over receive antennas
  - Inherent diversity over a static channel (constellation diversity) using single or multiple antenna(s)
    - Diversity improves with the number of constellation points
    - Unlike MIMO, diversity does not require sacrificing the rate
    - It essentially converts the Rayleigh fading channel into an AWGN channel with the same average receive energy and with a minor loss in capacity.



# Media-based: Rate



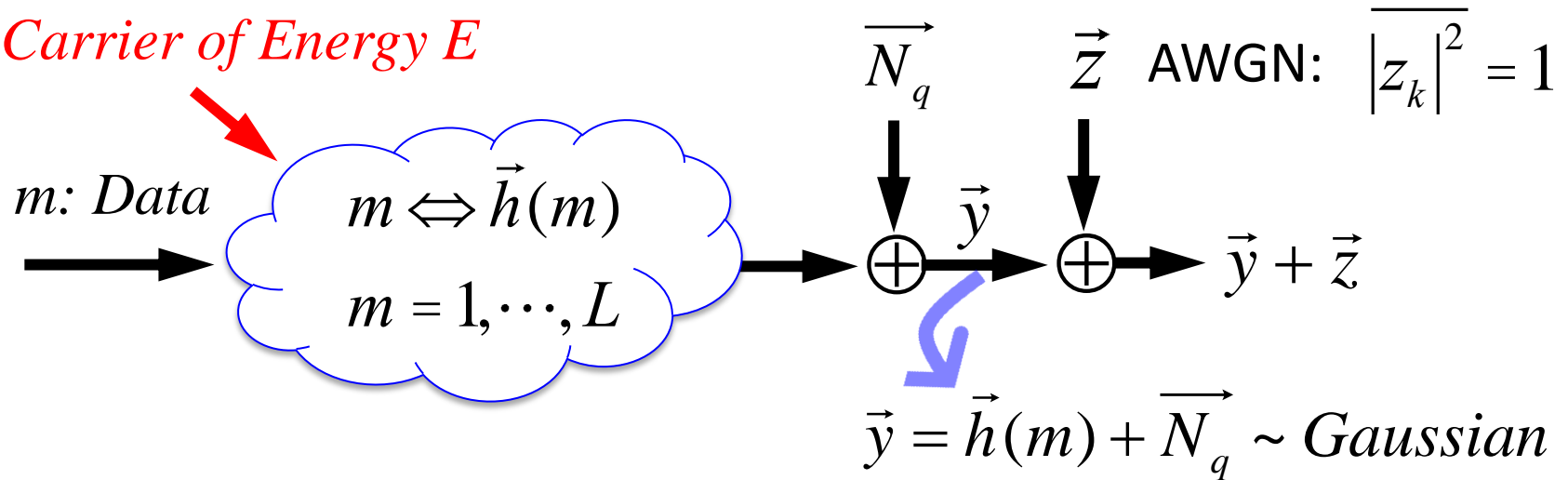
$$I(\vec{y}; m) = I(\vec{y}; \vec{h}(m)) = H(\vec{y}) - H(\vec{z}) = H(\vec{y}) - K \log_2(2\rho e S^2)$$

$\vec{h}(m), m = 1, \dots, L$  :  $K$ -D constellation (iid Gaussian elements)

# Gain due to Inherent Diversity:

## Typicality of Random Constellation

*Carrier of Energy E*



$$I = H(\vec{y}) - H(\vec{N}_q + \vec{z} | \vec{h}) \geq K \left[ \frac{1}{2} \log(2\pi e \sigma_Y^2) - E_{\vec{c}} \left\{ \frac{1}{2} \log 2\pi e (\sigma_N^2 + \sigma_{N_q | \vec{h}}^2) \right\} \right]$$

$$\geq K \left[ \frac{1}{2} \log(2\pi e \sigma_Y^2) - \int_{\vec{c} \in \mathcal{R}^Q} f_G(\vec{c}) \frac{1}{2} \log 2\pi e (\sigma_N^2 + \sigma_{N_q | \vec{h}}^2) d\vec{h} \right]$$

$$\sigma_{N_q | \vec{h}}^2 \leq \frac{L}{K} \int_{\vec{x} \in \mathcal{R}^2} f_G(\vec{x}) \left\| \vec{x} - \vec{h} \right\|^2 e^{-(L-1)P(\vec{x}, \vec{h})} d\vec{x}.$$



# Main Computational Tool

$$I \geq K \left[ \frac{1}{2} \log(2\pi e \sigma_Y^2) - E_{\vec{c}} \left\{ \frac{1}{2} \log 2\pi e (\sigma_N^2 + \sigma^2_{N_q|\vec{h}}) \right\} \right]$$

$$\geq K \left[ \frac{1}{2} \log(2\pi e \sigma_Y^2) - \int_{\vec{c} \in \mathfrak{R}^Q} f_G(\vec{h}) \frac{1}{2} \log 2\pi e (\sigma_N^2 + \sigma^2_{N_q|\vec{h}}) d\vec{h} \right]$$

$$\sigma^2_{N_q|\vec{h}} \leq \frac{L}{K} \int_{\vec{x} \in \mathfrak{R}^2} f_G(\vec{x}) \|\vec{x} - \vec{h}\|^2 e^{-(L-1)P(\vec{x}, \vec{h})} d\vec{x}$$

$$\cong \frac{2\Gamma(2/K + 1)}{K} \left( \frac{\Gamma(K/2 + 1)}{L} \right)^{\frac{2}{K}} e^{\frac{c^2}{Q}} = A e^{\frac{c^2}{K}} \left( \frac{1}{L} \right)^{2/K}$$

where,  $A = \frac{2\Gamma(2/K + 1)(\Gamma(K/2 + 1))^{\frac{2}{K}}}{K}$

As a result,  $\sigma^2_{N_q|\vec{h}} \cong \left( \frac{1}{L} \right)^{2/K} \rightarrow 0$ , as  $L \rightarrow \infty$



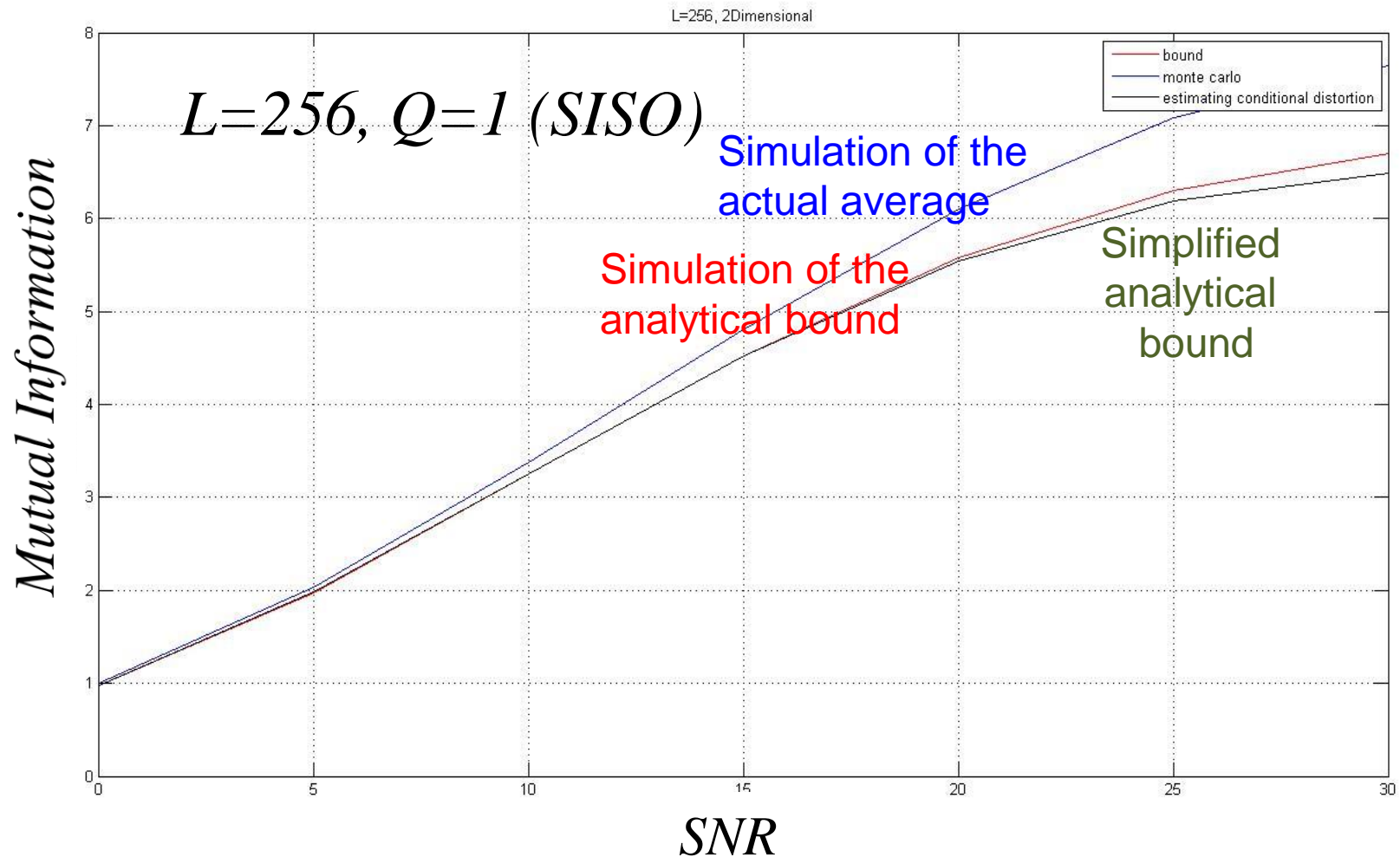


Main Conclusion of  $\sigma^2_{N_q|\bar{h}} \cong \left(\frac{1}{L}\right)^{2/K} \rightarrow 0, \text{ as } L \rightarrow \infty$

- Consider a slow Rayleigh fading channel for which statistical average of the fading gain per receive antenna is one.
  - Using a single TX and  $Q$  RX antennas over such channel, mutual information averaged over different realizations of a constellation with  $L$  points approaches the capacity of  $2Q$  parallel AWGN channels, each with unit energy, as  $L \rightarrow \infty$ .

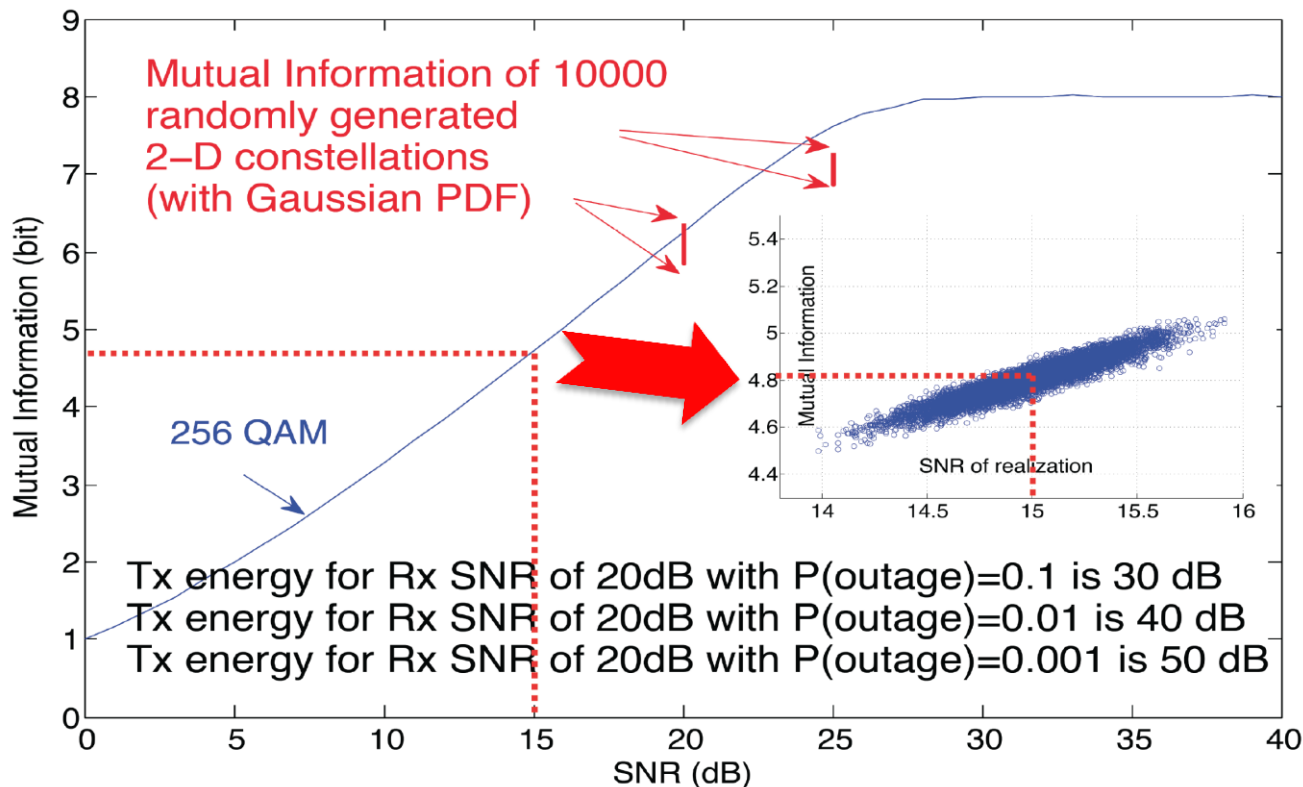


# Accuracy of the Computational Tool and its Simplified Version in Non-asymptotic Situations

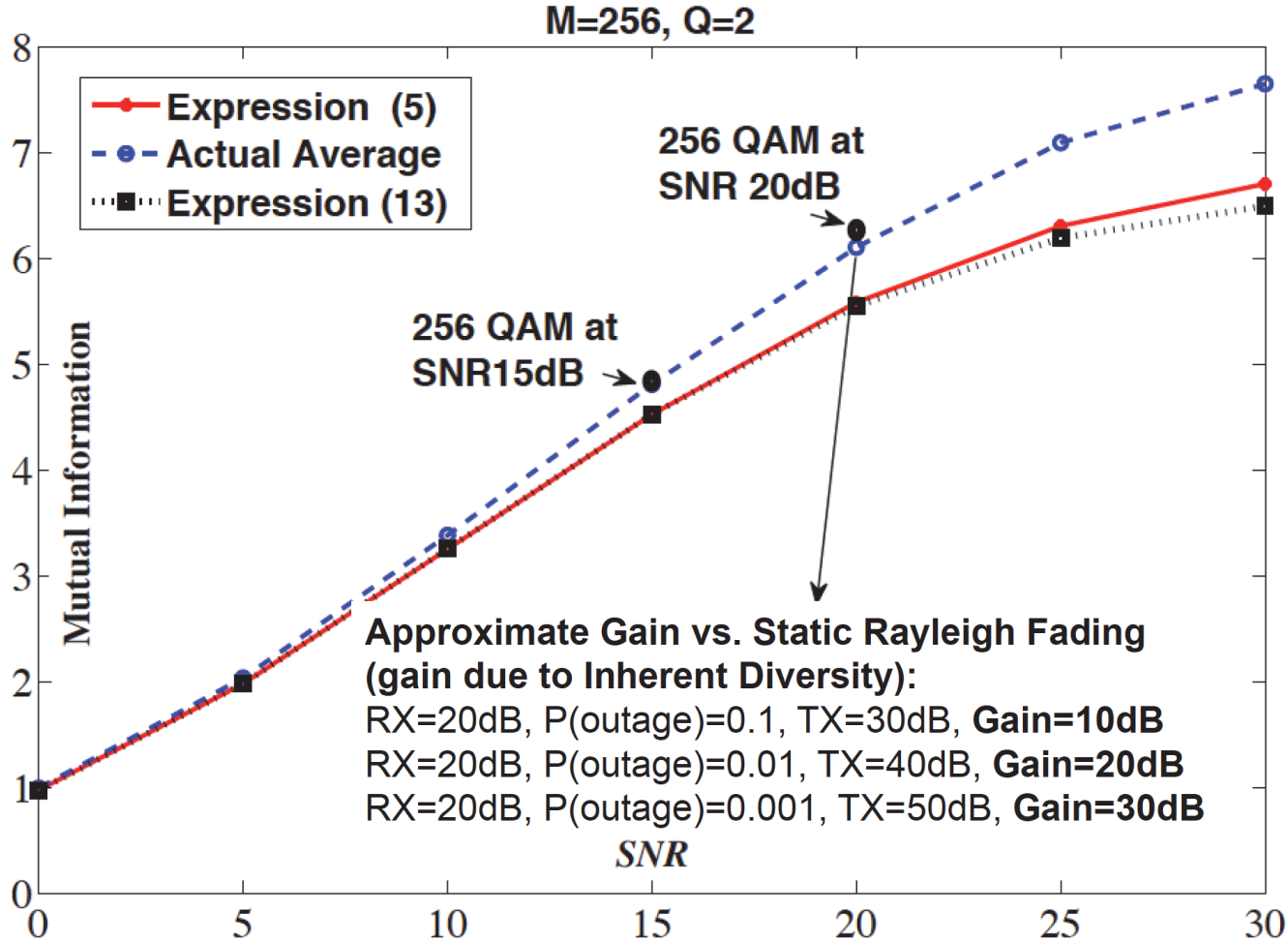


# Main Benefit: Inherent Diversity in A Single Constellation

- Conventional methods suffer from deep fades in slow fading.
- This problem disappears as “Good and Bad” channel realizations contribute to forming the constellation.

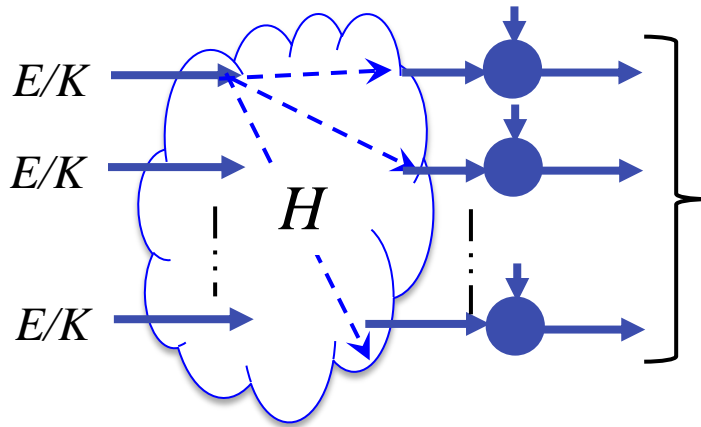


# Comparisons for the Average Rate



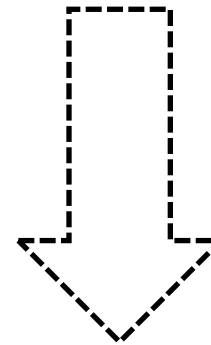
# Media-based vs. Source-based

$K \times K$   
MIMO  
 $K$  complex  
Dimensions

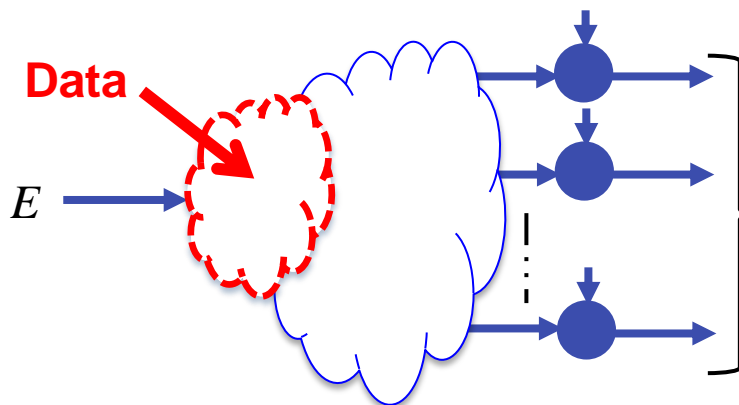


Total signal energy:  $KE$   
Basis: **Non-orthogonal**  
Complex Dimensions/sec/Hz:  $K$

**Better Performance**



Media-based  
**one complex  
dimension**

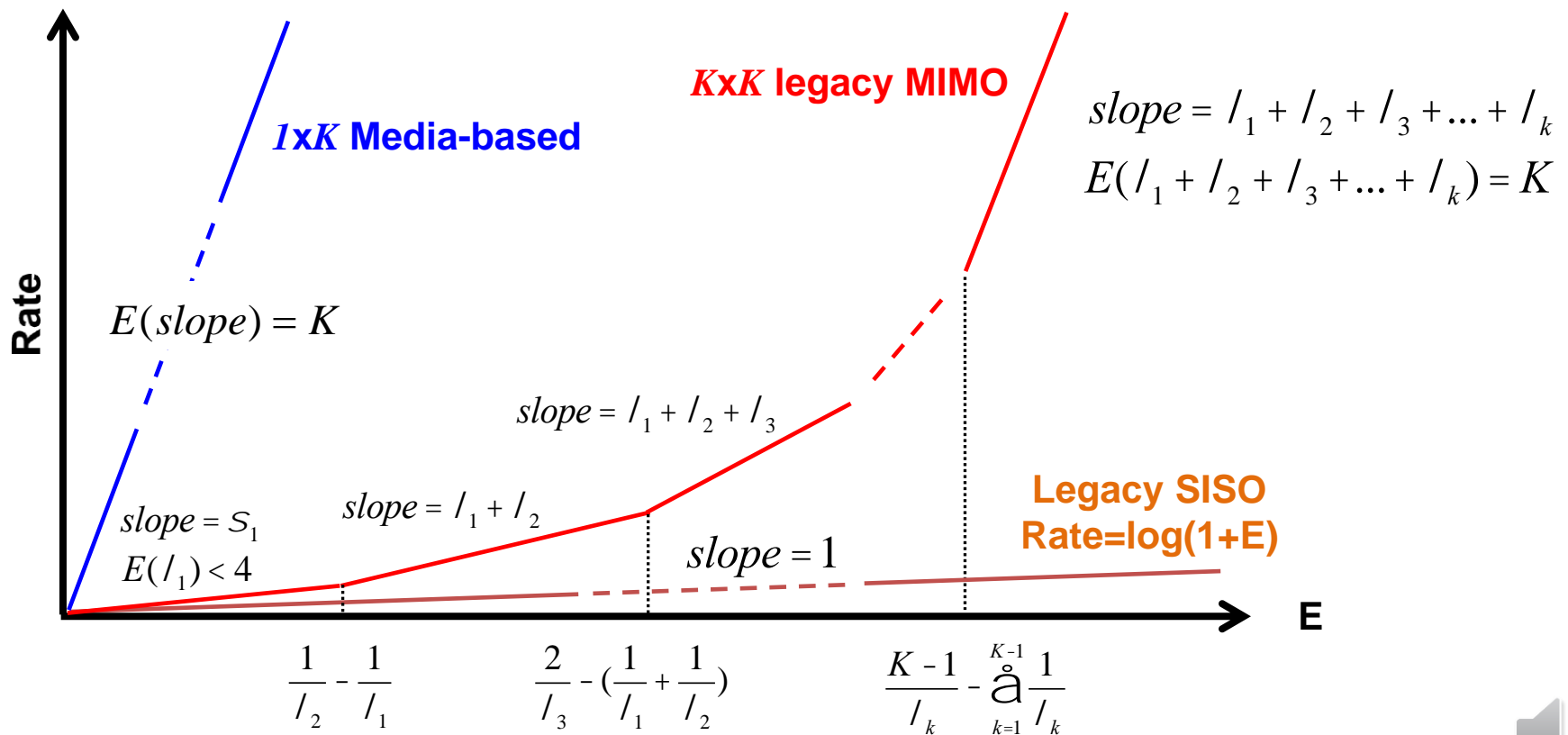


Total signal energy:  $KE$   
Basis: **Orthogonal**  
Complex Dimension/sec/Hz:  $K$



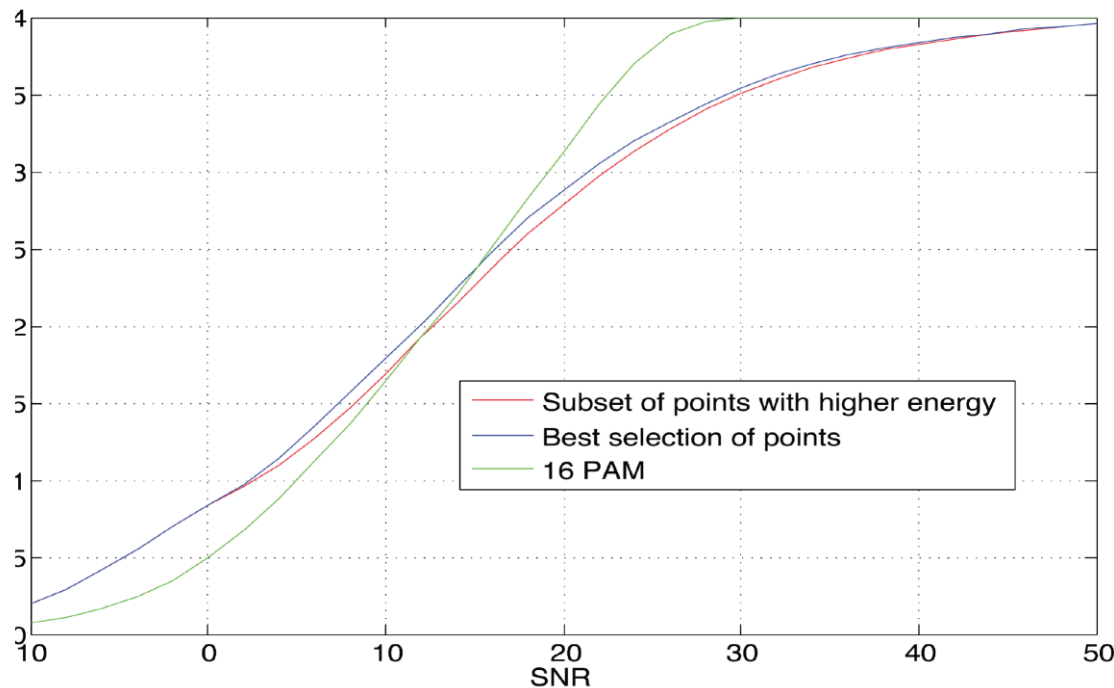
# Media-based vs. Legacy Systems: Effective Dimensionality

$\lambda_1 > \lambda_2 > \dots > \lambda_K$  : Eigenvalues of a  $K \times K$  Wishart random matrix



# Selection Gain

- Select a subset of points, which, subject to uniform probabilities, maximize the mutual information.
- In practice, using the subset of points with highest energy, which **maximizes the slope the rate at zero SNR**, performs very well.



# Media-based vs. Legacy Systems: Slope of Rate vs. SNR (dB) at SNR=0

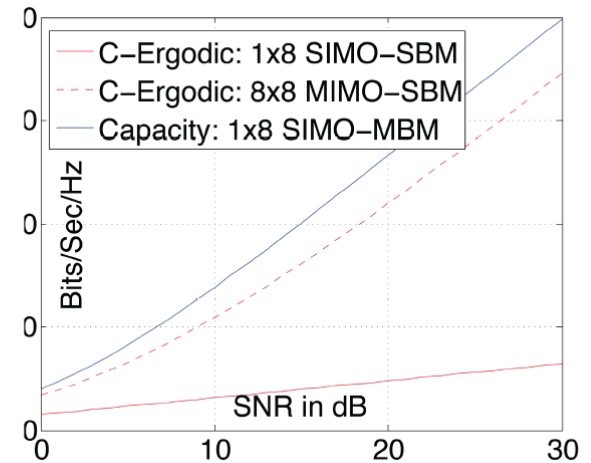
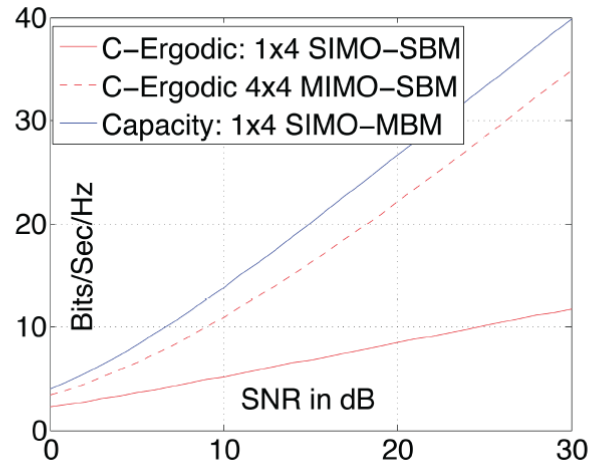
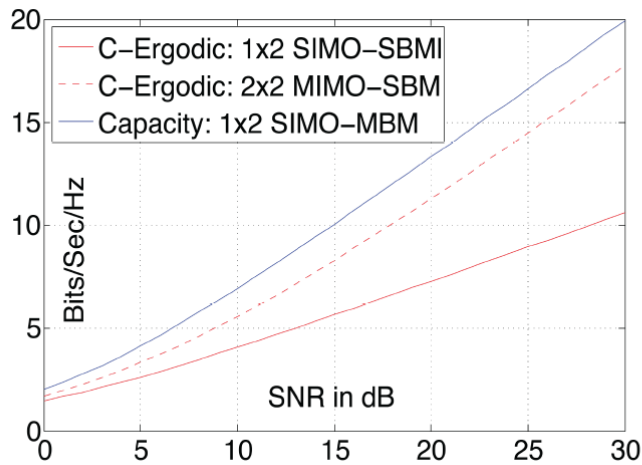
- Legacy SISO: Slope=1
- Legacy  $K \times K$  MIMO: **Maximum eigenvalue of a  $K \times K$  Wishart matrix (upper limited by 4)**
- $1 \times K$  Media-based:  $K$

	$L=1, K=2$	$L=1, K=4$	$L=1, K=8$	$L=1, K=infinity$
$K \times K$ MIMO	1.75	2.45	2.96	4
$1 \times K$ Media-based	2	4	8	Infinity

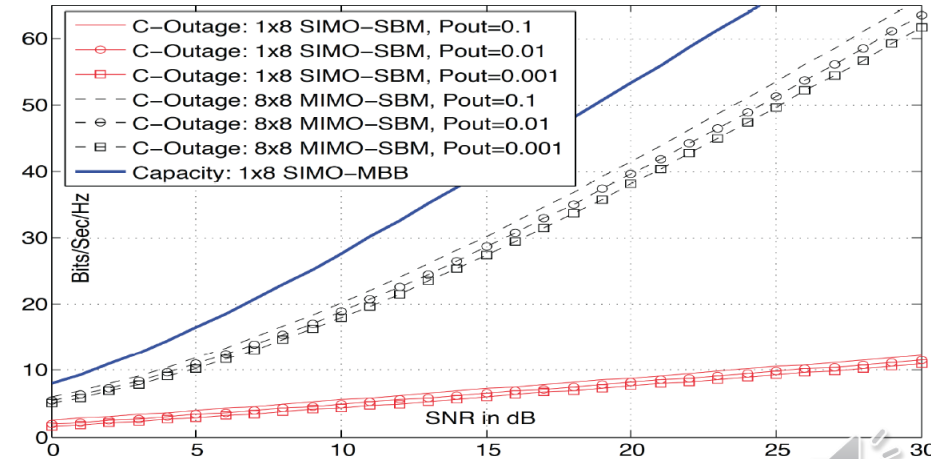
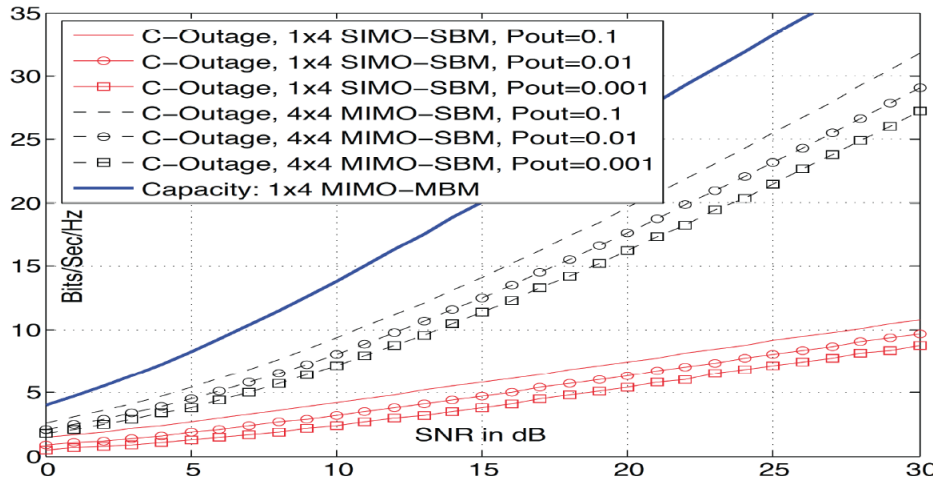
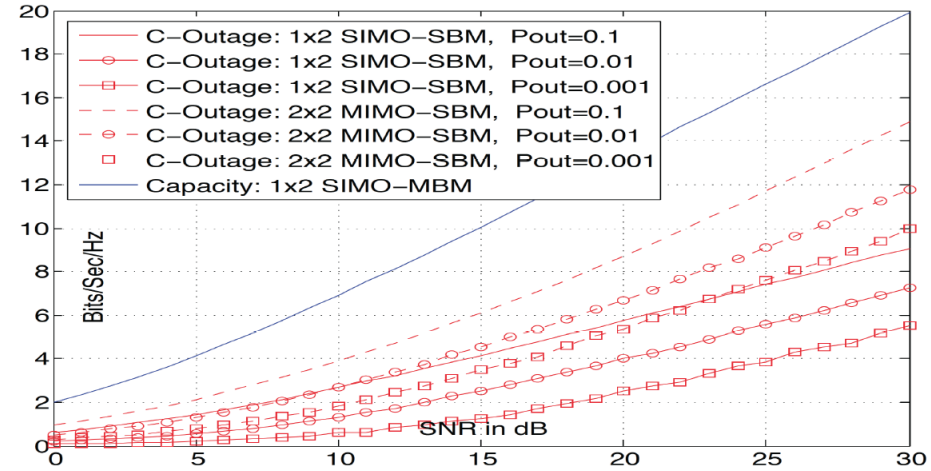
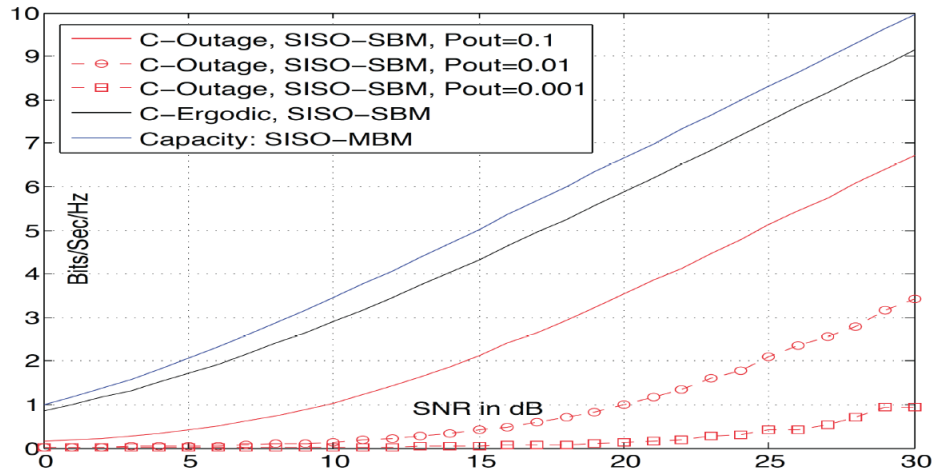
- **Selection Gain** further increases the slope of media-based to:  $\max \|\vec{c}_i\|, i = 1, \dots, L$ .
  - e.g. average slope scales as  $\log(L)$  for SISO case.



# Comparison with Ergodic Capacity



# Comparison with Outage Capacity



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